

FOURIER ANALYSIS

Reference

2

- <http://www.gaussianwaves.com/2015/11/interpreting-fft-results-complex-dft-frequency-bins-and-fftshift/>

Fourier analysis

3

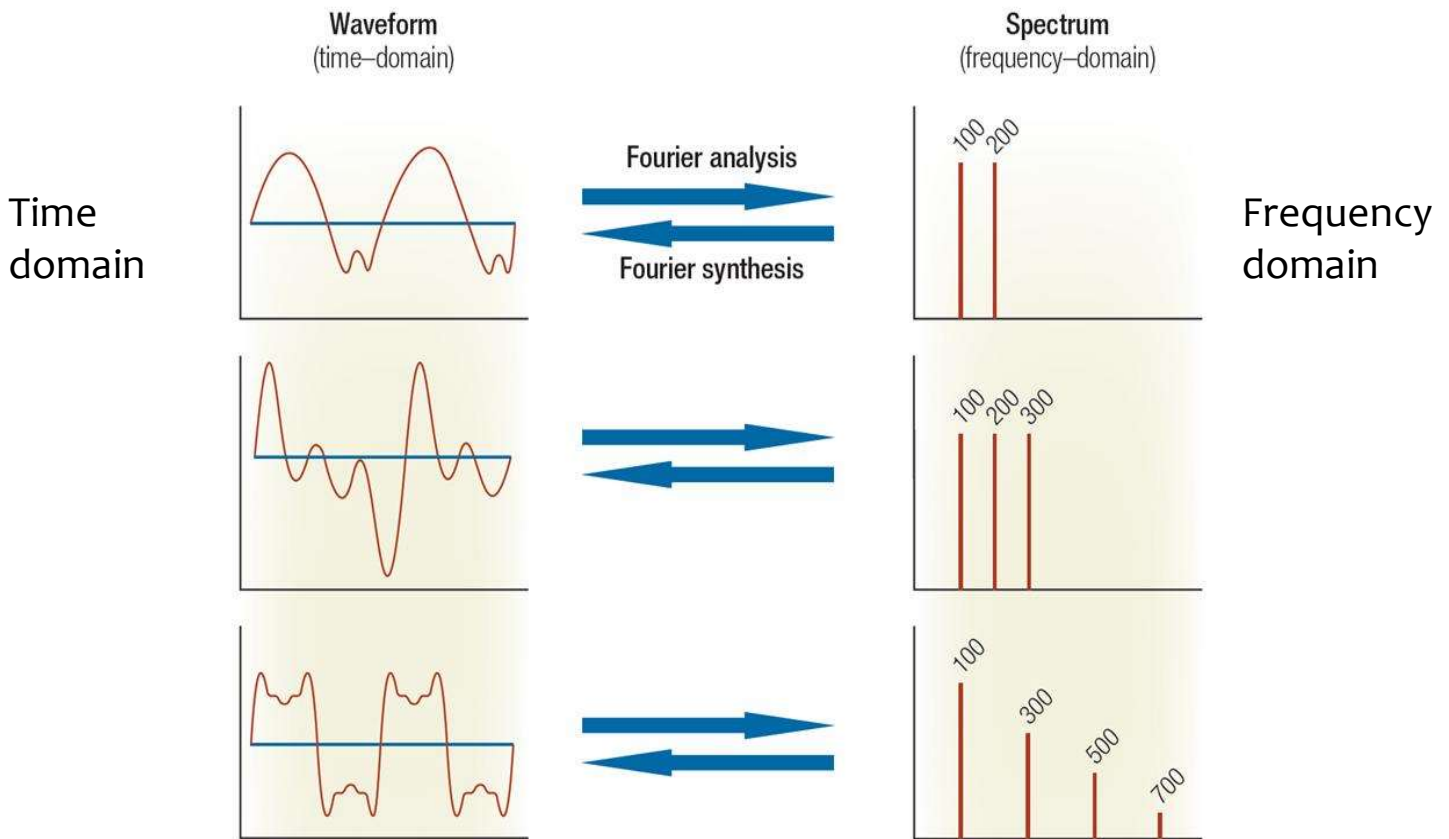
- Any complex, 1-dimensional function can be expressed as an additive series of sinusoidal functions varying in (1) frequency, (2) amplitude and (3) phase.
- (Continuous) Fourier transform

$$S(f) = \int_{-\infty}^{\infty} s(t) \cdot e^{-i2\pi ft} dt.$$

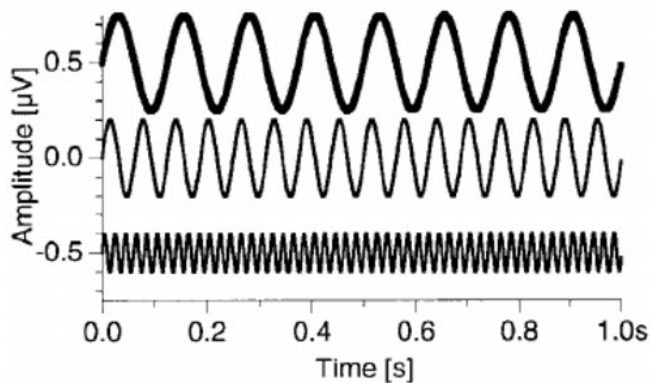
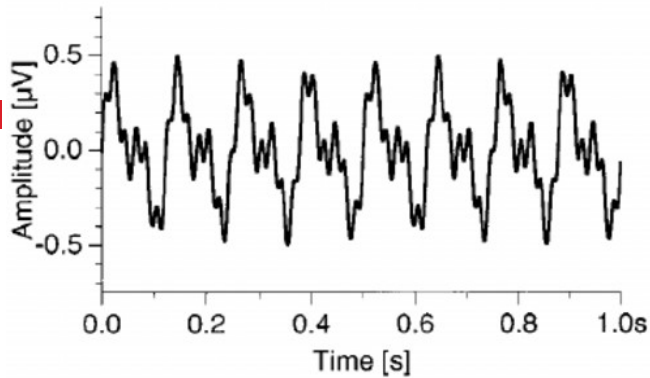
Euler's formula
 $e^{ix} = \cos x + i \sin x$

frequency-domain function

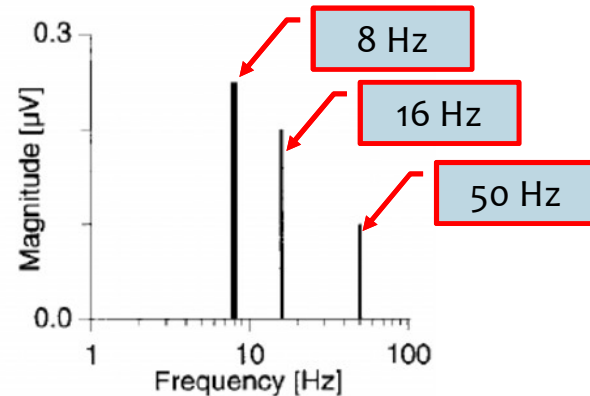
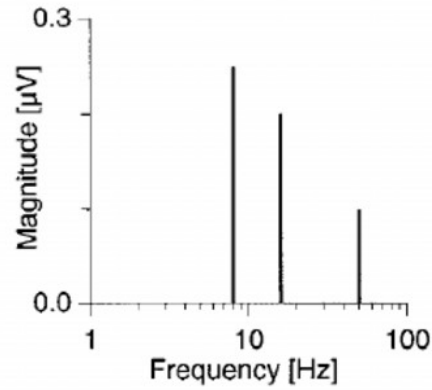
time-domain function



Time series



Spectrum



Fourier analysis

- The **top part** shows a somewhat irregular waveform with both slow and fast oscillations.
- The **bottom part** shows the three sinusoidal waveforms which, when added together, produce the top trace.
- The **lowest frequency** (thick trace) contains exactly **8** periods in the recording interval (=analysis interval) of 1 s length. Thus the corresponding spectral line (right) is located at **8 Hz**.
- The spectrum further reveals the second frequency of **16 Hz** and a third **50 Hz** component.

Four types of Fourier Transforms

7

- In signal processing , a time domain signal can be continuous or discrete and it can be aperiodic or periodic.
- This gives rise to four types of Fourier transforms.

Four types of Fourier Transforms

8

Transform	Nature of time domain signal	Nature of frequency spectrum
Fourier Transform (FT), (a.k.a Continuous Time Fourier Transform (CTFT))	continuous, non-periodic	continuous, non-periodic
Discrete-time Fourier Transform (DTFT)	discrete, non-periodic	continuous, periodic
Fourier Series (FS)	continuous, periodic	discrete, non-periodic
Discrete Fourier Transform (DFT)	discrete, periodic	discrete, periodic

Four types of Fourier Transforms

9

- We will limit our discussion to DFT, that is widely available as part of software packages like Matlab, Scipy(python) etc., however we can approximate other transforms using DFT.
- The DFT can be computed efficiently in practice using a fast Fourier transform (FFT) algorithm.

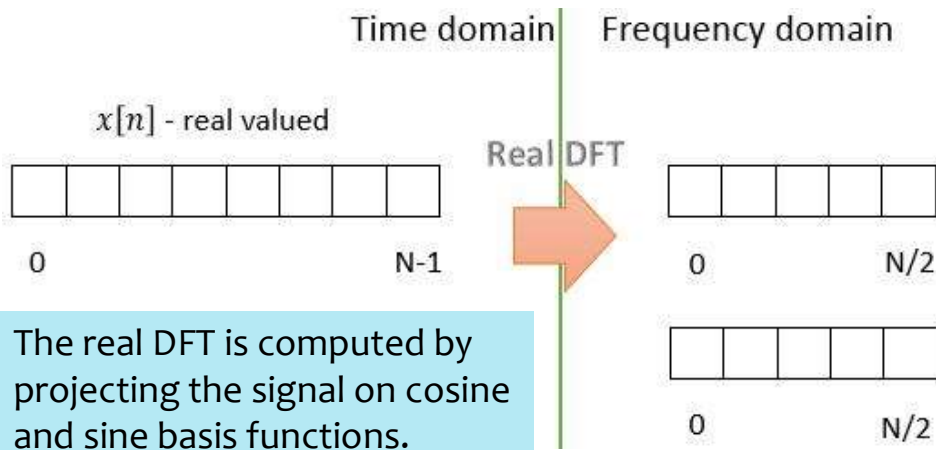
Real version and Complex version

10

- For each of the listed transforms above, there exist a real version and complex version.
- The real version of the transform, takes in a real numbers and gives two sets of real frequency domain points
 - ▣ one set representing coefficients over **cosine basis** function
 - ▣ and the other set representing the coefficients over **sine basis** function.
- The complex version of the transforms represent positive and negative frequencies in a single array.
 - ▣ The complex versions are flexible that it can process both complex valued signals and real valued signals.

Real DFT

11



The real DFT is computed by projecting the signal on cosine and sine basis functions.

$$X_{re}[k] = \frac{2}{N} \sum_{n=0}^{N-1} x[n] \cos\left(\frac{2\pi kn}{N}\right)$$

$X_{re}[k]$ – real valued (cosine terms)

$X_{im}[k]$ – real valued (sine terms)

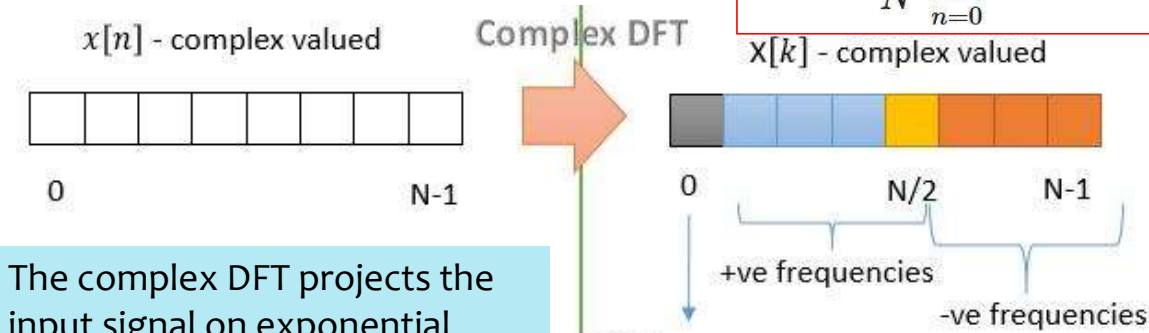
$$X_{im}[k] = -\frac{2}{N} \sum_{n=0}^{N-1} x[n] \sin\left(\frac{2\pi kn}{N}\right)$$

Complex DFT

12

Euler's formula
 $e^{ix} = \cos x + i \sin x$

$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j2\pi kn/N}$$



The complex DFT projects the input signal on exponential basis functions.

DC frequency component Nyquist frequency component at $(N/2)$ is common to positive and negative frequencies

Complex DFT

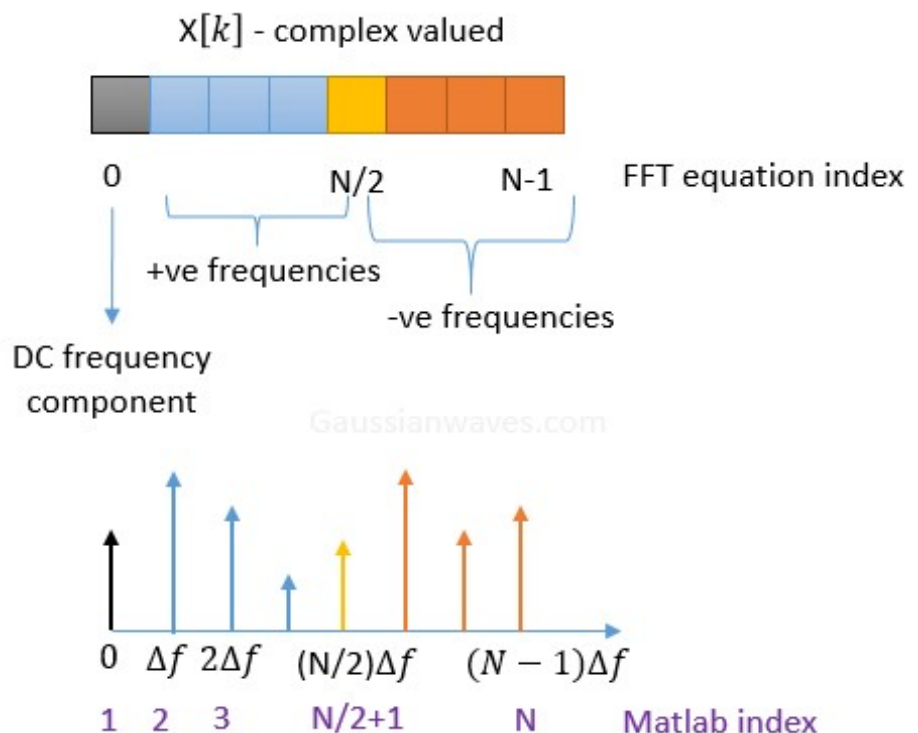
13

- The arrays values are interpreted as follows
 - ▣ $X[0]$ represents DC frequency component
 - ▣ Next $N/2$ terms are **positive** frequency components with $X[N/2]$ being the Nyquist frequency (which is equal to half of sampling frequency)
 - ▣ Next $N/2 - 1$ terms are **negative** frequency components
 - note: negative frequency components are the phasors rotating in opposite direction, they can be **optionally omitted** depending on the application

Complex DFT

14

- FFT is widely available in software packages like Matlab, Scipy etc...
- FFT in Matlab/Scipy implements the complex version of DFT.



Generate a cosine signal

15

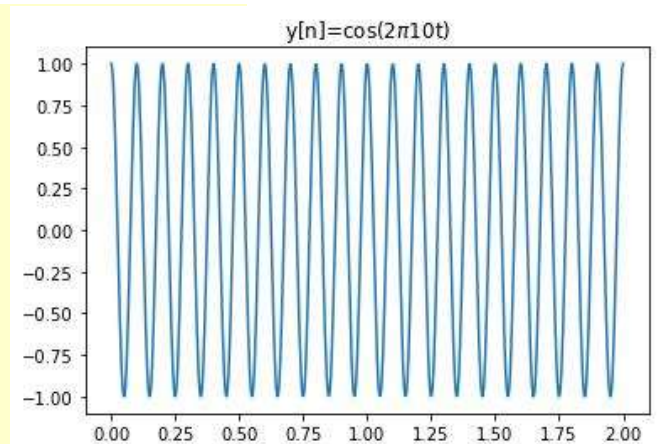
- Lets assume that the $y[n]$ is the time domain cosine signal of frequency $f_c=10\text{Hz}$ that is sampled at a frequency $f_s=32*f_c$ for representing it in the computer memory.

$$y[n] = \cos(2\pi f_c t)$$

Generate a cosine signal

16

```
import matplotlib.pyplot as plt
import numpy as np
# %matplotlib inline
# frequency of the carrier
fc = 10
# sampling frequency factor=32
fs = 32 * fc
duration = 2 # 2 seconds duration
t = np.linspace(0, duration, duration*fs)
# time domain signal (real number)
y = np.cos(2*np.pi*fc*t)
plt.title(r'y[n]=cos(2\pi$10t)')
plt.plot(t, y)
```



Interpreting the FFT results

17

- Compute the one-dimensional discrete Fourier Transform.

```
numpy.fft.fft(a, n=None, axis=-1, norm=None)
```
- Parameters:
 - ▣ `a` : array_like, Input array, can be complex.
 - ▣ `n` : int, optional, Length of the transformed axis of the output.
 - If `n` is smaller than the length of the input, the input is truncated .
 - If it is larger, the input is padded with zeros.
 - If `n` is not given, the length of the input along the axis specified by `axis` is used.
- Returns a complex ndarray

Interpreting the FFT results

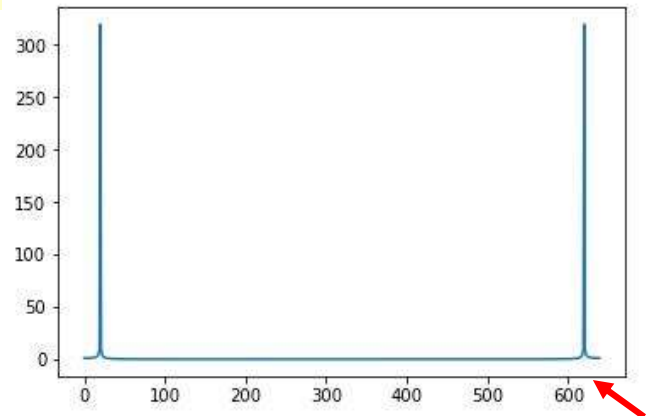
18

- Let's consider taking a $N=256$ point FFT, which is the 8th power of 2.
 - ▣ FFT length is generally considered as power of 2 – this is called radix-2.
- Note:
 - ▣ In our case, the cosine wave is of 2 seconds duration and it will have 640 points
 - A 10Hz frequency wave sampled at 32 times oversampling factor will have $2 \times 32 \times 10 = 640$ samples in 2 seconds of the record.
 - ▣ Since our input signal is periodic, we can safely use $N=256$ point FFT, anyways the FFT will extend the signal when computing the FFT .

Interpreting the FFT results

19

```
freqY = np.fft.fft(y)
spectrum = np.sqrt(freqY.real**2+freqY.imag**2)
plt.plot(spectrum)
```

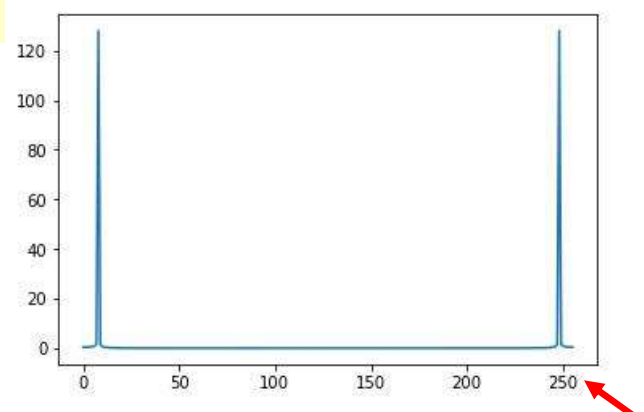


Interpreting the FFT results

20

```
N=256 #FFT size
freqY = np.fft.fft(y, N)
spectrum = np.sqrt(freqY.real**2+freqY.imag**2)
plt.plot(spectrum)
```

- Note that the index for the raw FFT are integers from $0 \rightarrow N-1$
- We need to convert the integer indices to frequencies.



Frequency axis transform

21

- `numpy.fft.fftfreq(n, d=1.0)`
 - The returned float array `f` contains the frequency bin centers in cycles per unit of the sample spacing (with zero at the start).
 - For instance, if the sample spacing is in seconds, then the frequency unit is cycles/second.

```
>>> signal = np.array([-2, 8, 6, 4, 1, 0, 3, 5], dtype=float)
>>> fourier = np.fft.fft(signal)
>>> n = signal.size
>>> timestep = 0.1
>>> freq = np.fft.fftfreq(n, d=timestep)
>>> freq
array([ 0. ,  1.25,  2.5 ,  3.75, -5.  , -3.75, -2.5 , -1.25])
```

Frequency axis transform

22

- frequency signal $f_c=10\text{Hz}$, sampling frequency $f_s=32*f_c$
- One second has $10*32=320$ samples
- sample spacing in second = $1/320$

```
In [86]: freq = np.fft.fftfreq(N, d=1/320)
In [87]: freq[8]
Out[87]: 10.0
In [89]: freq
Out[89]: array([ 0. ,  1.25,  2.5 , ..., -3.75, -2.5 , -1.25])
```

Frequency axis transform

23

- The cosine signal has a peak at 10Hz. In addition to that, it has also a peak at $256-8=248$ th sample that belongs to negative frequency portion.

```
spectrum[8]
Out[62]: 128.06700088210565
spectrum[248]
Out[65]: 128.06700088210559
```

```
In [91]: freq[8]
Out[91]: 10.0
In [92]: freq[248]
Out[92]: -10.0
```

The 10Hz cosine signal will leave a peak at the 8th sample ($10/1.25=8$)

$$\Delta f = \frac{f_s}{N} = \frac{32 * f_c}{256} = \frac{320}{256} = 1.25Hz$$

Frequency axis transform

24

- The sample at the Nyquist frequency ($f_s/2$) mark the boundary between the positive and negative frequencies.

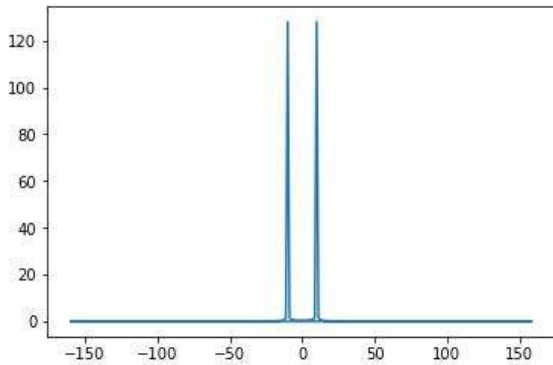
```
In [89]: freq
Out[89]: array([ 0. , 1.25, 2.5 , ..., -3.75, -2.5 , -1.25])
In [100]: nyquistIndex=int(N/2)
In [101]: freq[nyquistIndex]
Out[101]: -160.0
In [102]: freq[nyquistIndex-1]
Out[102]: 158.75
```

FFTShift

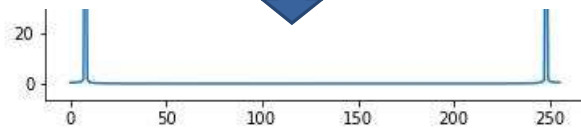
25

```
plt.plot(freq, spectrum)
```

Plot x, y pair
The sequence order is not changed.



```
In [118]: freq
Out[118]:
[ 0.    1.25  2.5   ..., -3.75 -2.5  -1.25]
In [119]: spectrum[:10]
Out[119]:
array([ 0.39984513,  0.40617256,  0.42641611,
        0.46504507,  0.53258955,  0.65487813,  0.91034296,
        1.6889461,  128.06700088,  1.52828374])
```



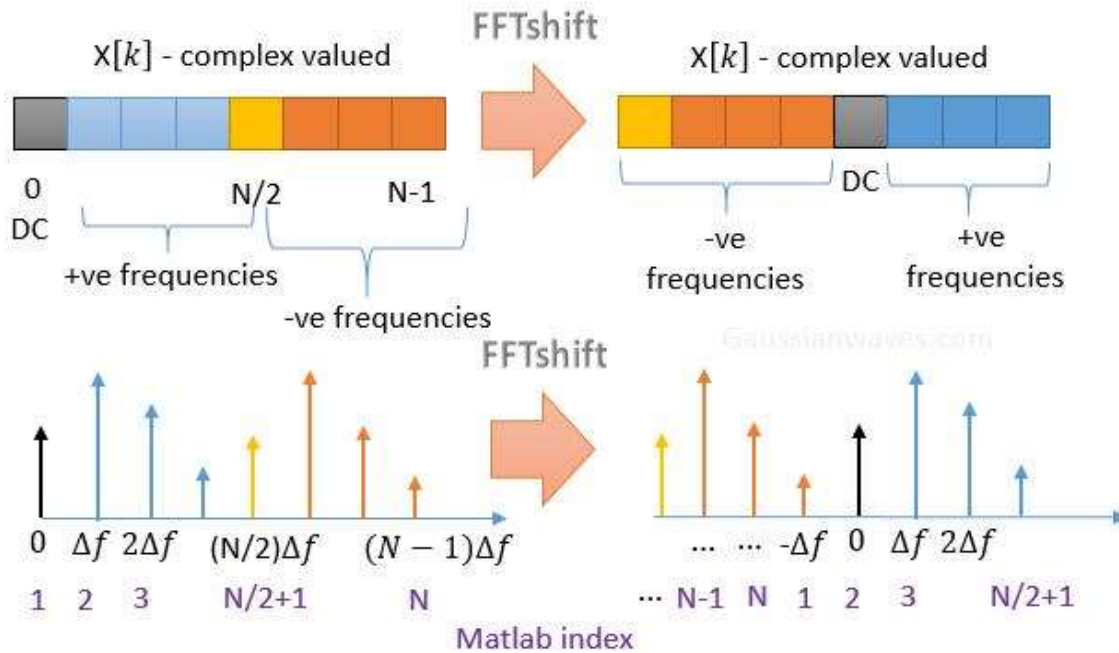
FFTShift

26

- From the plot we see that the frequency axis starts with DC, followed by positive frequency terms which is in turn followed by the negative frequency terms.
- To introduce proper order in the x-axis, one can use FFTshift function, which arranges the frequencies in order:
 - ▣ negative frequencies → DC → positive frequencies.
 - ▣ The fftshift function need to be carefully used **when N is odd**.

FFTShift

27



FFTShift

28

- **numpy.fft.fftshift(x, axes=None)**
 - ▣ Shift the zero-frequency component to the center of the spectrum.
 - ▣ Note that $y[0]$ is the Nyquist component only if $\text{len}(x)$ is even.
- **numpy.fft.ifftshift(x, axes=None)**
 - ▣ The inverse of `fftshift`.
 - ▣ Although identical for even-length x , the functions differ by one sample for odd-length x .

```
>>> freqs = np.fft.fftfreq(10, 0.1)
>>> freqs
array([ 0.,  1.,  2.,  3.,  4., -5., -4., -3., -2., -1.])
>>> np.fft.fftshift(freqs)
array([-5., -4., -3., -2., -1.,  0.,  1.,  2.,  3.,  4.])
```

FFTShift

29

```
shift_freq = np.fft.fftshift(freq)
shift_spec = np.fft.fftshift(spectrum)
plt.figure()
plt.plot(shift_freq, shift_spec)
```

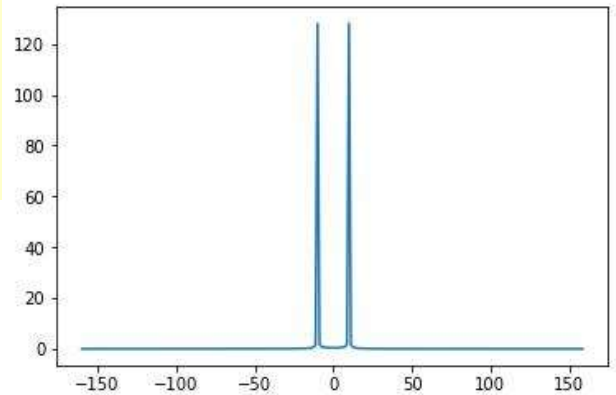
```
shift_freq
```

```
Out[118]: array([-160.   , -158.75, -157.5 , ..., 156.25, 157.5 , 158.75])
```

```
In [120]: shift_spec[:10]
```

```
Out[120]:
```

```
array([ 0.00232974, 0.00233041,
 0.00233242, 0.00233576, 0.00234045,
 0.00234648, 0.00235386, 0.0023626 ,
 0.0023727 , 0.00238418])
```



Frequency filtering

30

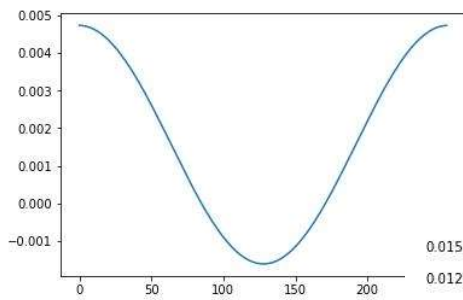
- filtering by setting cut-off frequency

```
lowPassMask = abs(freq) <=2 # cut-off frequency=2
print('non_zero= ', np.count_nonzero(lowPassMask))
lowPassFy = freqY.copy()
lowPassFy[~lowPassMask] = 0 # ~, equivalent to logical_not
lowPassY = np.fft.ifft(lowPassFy)
plt.figure()
plt.plot(lowPassY.real)
```

ifft() output is complex values. We only get the real part.

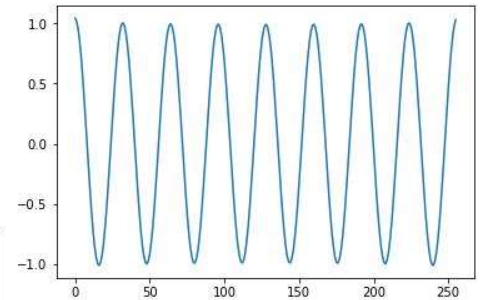
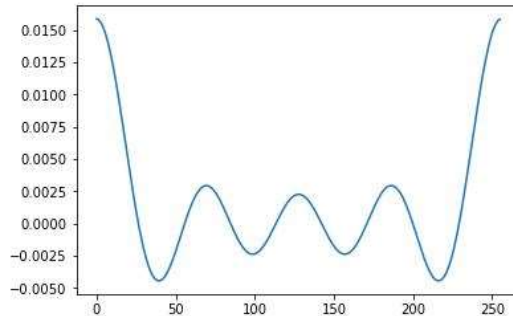
Low-pass filter

31



cut-off frequency = 2

cut-off frequency = 5



cut-off frequency = 10

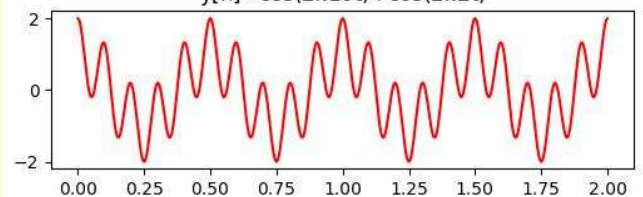
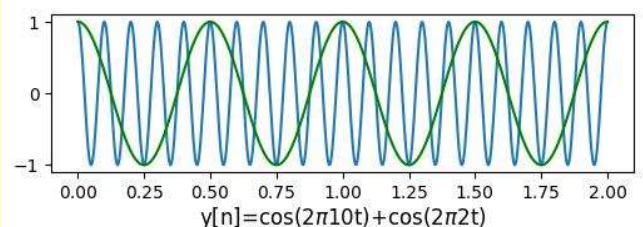
Example: Different frequency components in a signal

32

```
fs = 320 # sampling frequency
duration = 2 # 2 seconds duration
t = np.linspace(0, duration, duration*fs)
y10 = np.cos(2*np.pi*10*t)
y2 = np.cos(2*np.pi*2*t)
y = y10 + y2
```

```
plt.figure()
f, (ax1, ax2) = plt.subplots(2, 1)
plt.subplots_adjust(hspace = 0.4)
ax1.plot(t, y10)
ax1.plot(t, y2, 'g')
ax2.set_title(r'y[n]=cos(2$\pi$10t)+cos(2$\pi$2t)')
ax2.plot(t, y, 'r')
```

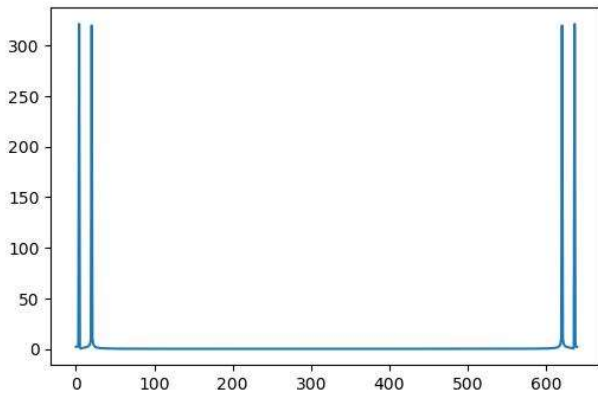
Signal: $y(t) = \cos(2\pi \cdot 10 \cdot t) + \cos(2\pi \cdot 2 \cdot t)$



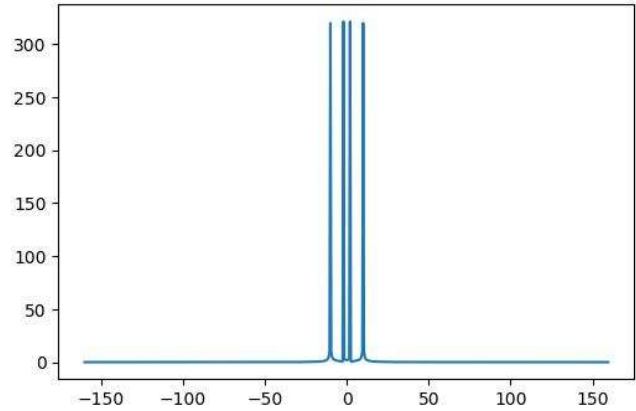
Many frequency components

33

- Fourier spectrum (N=640 #FFT size)



Before fftshift

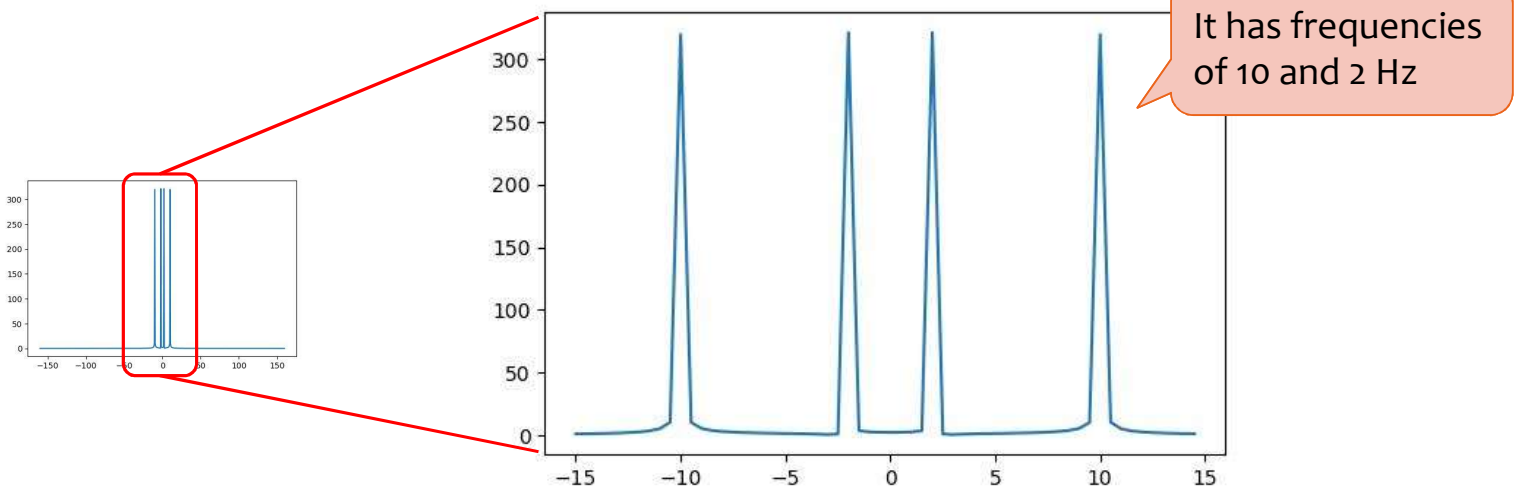


After fftshift

Many frequency components

34

- `plt.plot(shift_freq[290:350], shift_spec[290:350])`

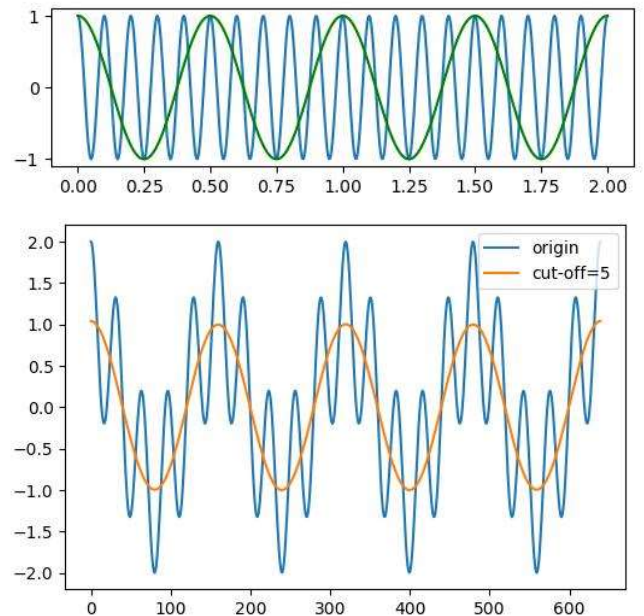


Low-pass

35

- cut-off frequency = 5

```
n = len(y)
freqY = np.fft.fft(y)
freq = np.fft.fftfreq(n, d=1/fs)
lowPassMask = abs(freq) <= 5
lowPassFy = freqY.copy()
lowPassFy[~lowPassMask] = 0
lowPassY = np.fft.ifft(lowPassFy)
plt.figure()
fig, ax = plt.subplots()
ax.plot(y, label='origin')
ax.plot(lowPassY.real, label='cut-off=5')
legend = ax.legend()
```

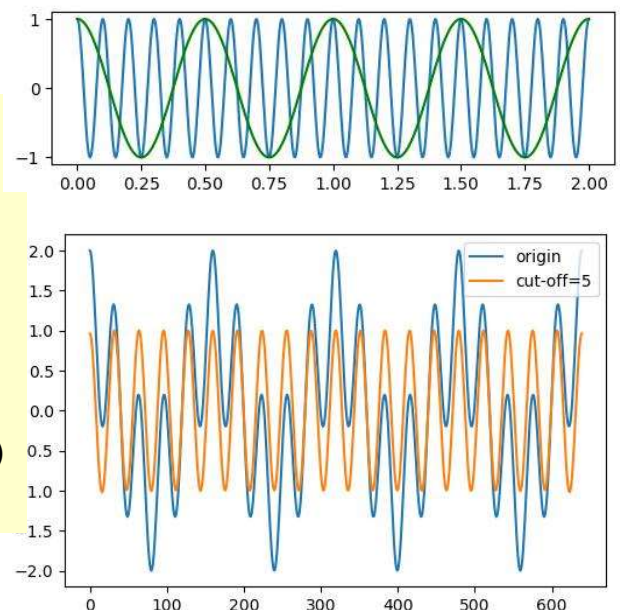


High-pass

36

- cut-off frequency = 5

```
highPassMask = abs(freq) >= 5
highPassFy = freqY.copy()
highPassFy[~highPassMask] = 0
highPassY = np.fft.ifft(highPassFy)
plt.figure()
fig, ax = plt.subplots()
ax.plot(y, label='origin')
ax.plot(highPassY.real, label='cut-off=5')
legend = ax.legend()
```



ALTERNATIVE METHOD FOR CREATING LOWPASS FILTER IN SCIPY

Reference

38

- <http://stackoverflow.com/questions/25191620/creating-lowpass-filter-in-scipy-understanding-methods-and-units>
- **Scipy Signal processing ([scipy.signal](https://docs.scipy.org/doc/scipy/reference/signal.html))**
 - <https://docs.scipy.org/doc/scipy/reference/signal.html>

Butterworth filter

39

- `scipy.signal.butter(N, Wn, btype='low', analog=False, output='ba')`
 - ▣ Butterworth digital and analog filter design.
 - ▣ Parameters
 - **N** : int, The order of the filter.
 - **Wn** : array_like, A scalar or length-2 sequence giving the critical frequencies.
 - **btype** : {'lowpass', 'highpass', 'bandpass', 'bandstop'}, optional
 - Default is 'lowpass'.

Butterworth filter

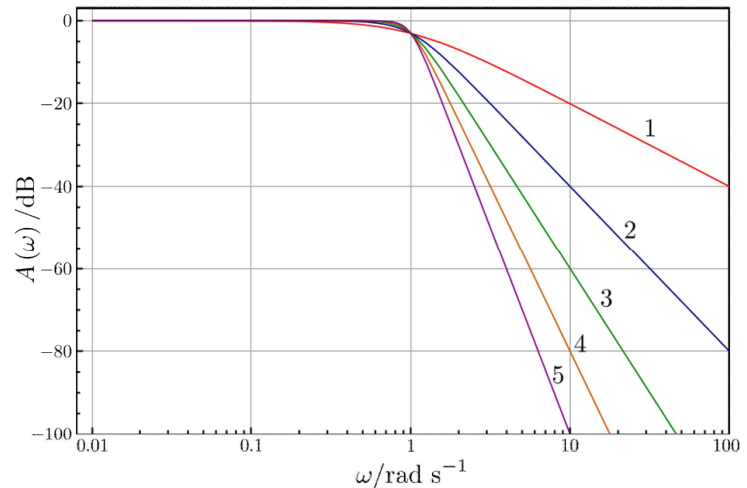
40

- **Parameter Wn**
 - For a Butterworth filter, this is the point at which the gain drops to $1/\sqrt{2}$ that of the passband (the “-3 dB point”).
 - For digital filters, Wn is normalized from 0 to 1, where 1 is the Nyquist frequency, π radians/sample. (Wn is thus in half-cycles / sample.)
 - For analog filters, Wn is an angular frequency (e.g. rad/s).
- **Returns**
 - ▣ **b, a** : ndarray, ndarray
 - Numerator (*b*) and denominator (*a*) polynomials of the IIR filter. Only returned if `output='ba'`.

Butterworth filter

41

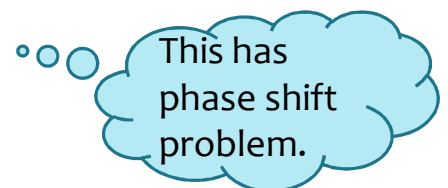
- Plot of the gain of Butterworth low-pass filters of orders 1 through 5, with cutoff frequency $\omega_0=1$.



scipy.signal.lfilter

42

- `scipy.signal.lfilter(b, a, x, axis=-1, zi=None)`
 - Filter a data sequence, x , using a digital filter.
 - Parameters
 - **b, a**: ndarray, ndarray
 - Numerator (b) and denominator (a) coefficient vectors in a 1-D sequence.
 - **x**: array_like, An N-dimensional input array.
 - Return
 - **y**: array, The output of the digital filter.
- Use `scipy.signal.filtfilt(b,a, x, ...)` instead of `lfilter()`
 - This function applies a linear filter twice, once forward and once backwards.



Creating lowpass filter

43

```
import scipy.signal as sg
def butter_lowpass(cutoff, fs, order=5):
    nyq = 0.5 * fs
    normal_cutoff = cutoff / nyq
    b, a = sg.butter(order, normal_cutoff, btype='low', analog=False)
    return b, a

def butter_lowpass_filter(data, cutoff, fs, order=5):
    b, a = butter_lowpass(cutoff, fs, order=order)
    y = sg.filtfilt(b, a, data)
    return y
```

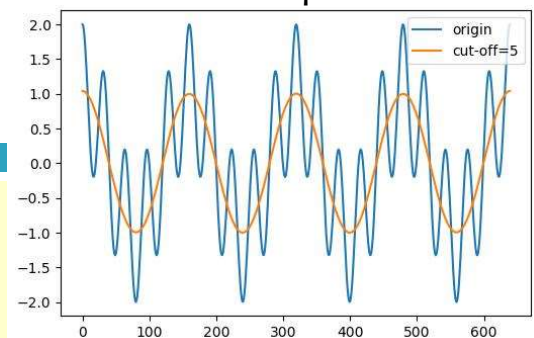
Creating lowpass filter

44

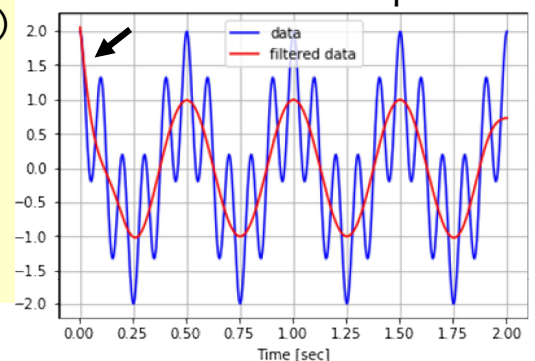
```
# Filter the data, and plot both the original
# and filtered signals.
order = 6
cutoff = 5

out = butter_lowpass_filter(y, cutoff, fs, order)
plt.figure()
plt.plot(t, y, 'b-', label='data')
plt.plot(t, out, 'g-', label='filtered data')
plt.xlabel('Time [sec]')
plt.grid()
plt.legend()
```

FFT low-pass



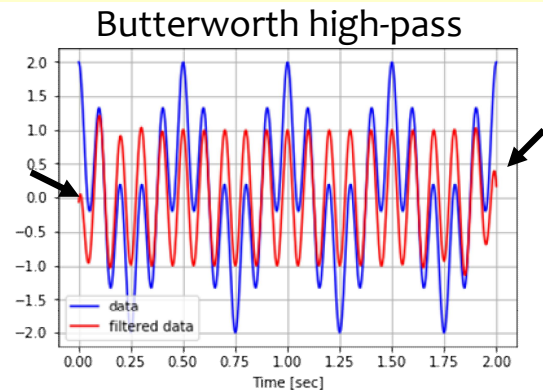
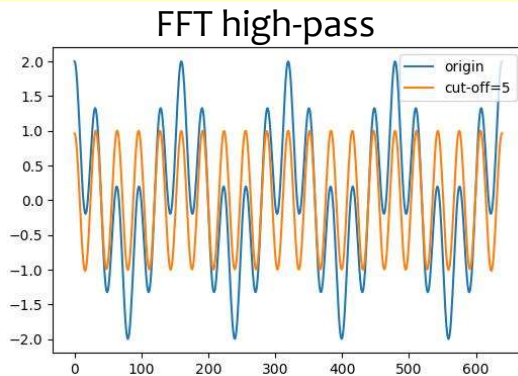
Butterworth low-pass



Creating highpass filter

45

```
def butter_highpass(cutoff, fs, order=5):  
    nyq = 0.5 * fs  
    normal_cutoff = cutoff / nyq  
    b, a = sg.butter(order, normal_cutoff, btype='high', analog=False)  
    return b, a
```



Computation time

46

- ❑ Butterworth filter is faster than FFT
- ❑ Measure the computation time

```
from time import time  
t1 = time()  
ax_low_buf = butter_lowpass_filter(ax, cutoff, fs)  
ax_high_buf = butter_highpass_filter(ax, cutoff, fs)  
t2 = time()  
print('Butterworth low high pass takes %f seconds for  
ax, ay, az' % (t2-t1))
```